

# **MAS6230**

# High Efficiency 60W Synchronous Buck-Boost DC/DC Converter

- Wide Input Voltage Range from 7V to 36V
- Adjustable Output Voltage from 4V to 20V
- Up to 60W Output Power
- Very High Efficiency >90%
- Up to 3A Output Current (Buck Mode)
- Over Temperature Protection
- Output Overvoltage Protection
- Low Quiescent Current
- QFN-48 7x7x0.75 mm Package

#### DESCRIPTION

MAS6230 is a power supply solution for cell phone, tablet and laptop cigarette lighter chargers in cars and other vehicles. It comprises of a very high efficiency synchronous buck-boost DC/DC converter which can deliver up to 60W output power. It has both wide 7V to 36V input and 4V to 20V output voltage range covering wide variety of applications. The DC/DC converter has hysteresis between buck and boost modes which makes it stable with wide range of external components.

MAS6230 integrates over temperature (OTP) and output overvoltage protection (OVP) circuitry to prevent damage under fault conditions. The output voltage is adjustable by external feedback resistors. Additionally the output voltage can be selected from four different options using feedback voltage selection pins (FBSEL0-1). This is a useful feature in applications where several different output voltage selections are supported.

The device is available in a thin 7x7x0.75 mm 48-lead QFN package.

#### FEATURES

- Wide input voltage range from 7V to 36V
- Output voltage adjustable by 2 external resistors from 4V to 20V
- Four output voltage options selectable by two input pins
- Up to 60W output power
- Over 90% efficiency
- Up to 3A output current (buck mode)
- Low inductance value inductor
- Automatic transition between buck and boost modes
- Over temperature protection
- Output overvoltage protection
- Low quiescent current
- Thin 7x7x0.75 mm QFN-48 package

#### APPLICATIONS

- USB-C PD car charger
- Power supply solution for industrial equipment
- Power supply solution for home appliances



# **ABSOLUTE MAXIMUM RATINGS**

			All voltage	es with respect t	o ground.
Parameter	Symbol	Conditions	Min	Мах	Unit
Supply Pin Voltage	VIN		-0.3	42	V
Output Pin Voltage	VOUT		-0.3	25	V
Internal Power Supply Pin Voltage	VDD		-0.3	6	V
Input Pin Voltage	EN		-0.3	VIN+0.3	V
	MEMO, FBSEL0-1, FB, FORCEBUCK		-0.3	6	V
Inductor Switch Pin	SW1		-0.3	42	V
Voltage	SW2		-0.3	25	V
Flying Capacitor Pin Voltage	FLY1A/B, FLY2A/B		-0.3	42	V
Storage Temperature	T <sub>STG</sub>		-40	+125	°C
ESD Voltage Rating	V <sub>HBM</sub>	Human Body Model (HBM)	±1		kV
	V <sub>CDM</sub>	Charged-Device Model (CDM)	±500		V

Note: Stresses beyond the values listed may cause a permanent damage to the device. The device may not operate under these conditions, but it will not be destroyed.

Note: JEDEC document JEP155 states that 500V HBM allows safe manufacturing with a standard ESD control process.

Note: JEDEC document JEP157 states that 250V CDM allows safe manufacturing with a standard ESD control process.

# **RECOMMENDED OPERATING CONDITIONS**

			A	ll voltages w	ith respect to	o ground.
Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Operating Junction Temperature	TJ		-40		+125	°C
Operating Ambient Temperature	TA		-40	+27	+85	°C
Operating Supply Voltage	V <sub>IN</sub>		7.0		36.0	V
Operating Output Voltage	Vout		5		20	V
Output Current	Ιουτ	Buck mode (V <sub>IN</sub> ≥ V <sub>OUT</sub> ) Note 1			3.0	А
	L	Shielded inductor Note 2	4	6.8	13	μH
Inductor Specification	RATED			6		А
	RDC			27		mΩ

**Note 1:** In boost mode the maximum output current may be lower due to coil current limiting. For further information see figure 4 on page 7. **Note 2:** See inductor examples in table 1 on page 6.



# **ELECTRICAL CHARACTERISTICS**

Demonster	T <sub>A</sub> =	-40°C to +85°C, typical values at $T_A = 27$ °C,	V <sub>IN</sub> = 20 V, I	_=6.8 μH; ur	less otherw	ise specifi
Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Input voltage range	VIN		7		36	V
Output voltage range	Vout		4		20.5	V
Internal supply voltage <sup>(1)</sup>	Vdd	VIN=20V	3.0	4.5	5.5	V
Efficiency	η	Iout=3A, VIN=30V, Vout=20V (buck mode)		92.4		%
Feedback voltage	Vfb	V <sub>IN</sub> =20V FBSEL1/0=LOW/LOW FBSEL1/0=LOW/HIGH FBSEL1/0=HIGH/LOW FBSEL1/0=HIGH/HIGH	1170 877 526 291	1200 900 540 300	1230 923 554 315	mV
Valley coil current limit	ICOIL_LIM	L=6.8 μH	3.3	3.8	4.3	Α
Peak coil current limit	I <sub>COIL_PEAK</sub>	L=6.8 μH		5.3	6.3	Α
On resistances of High side switches	RHON1 RHON2			30 50		mΩ
On resistances of Low side switches	Rlon1 Rlon2			50 50		mΩ
Shutdown current	Isd	EN=LOW, VOUT=0V		TBD		μA
Undervoltage lockout threshold voltage	Vuvlo+ Vuvlo-	V <sub>IN</sub> rising V <sub>IN</sub> falling	6.2 5.85	6.6 6.3	7 6.65	V
Undervoltage lockout hysteresis	VUVLOH	V <sub>IN</sub> rising		300		mV
Input high voltage	Vih	Pins EN, FBSEL0-1, FORCEBUCK	1.2			V
Input low voltage	VIL	Pins EN, FBSEL0-1, FORCEBUCK			0.4	V
Input current	IIH	Pins EN, FBSEL0-1, FORCEBUCK Input voltage = 3.5V, VIN = 20V		0.01	1	μA
	IIL <sup>(2)</sup>	Pins EN, FBSEL0-1, FORCEBUCK, VIN = 20V Input voltage = 0V Input voltage = 0.4V		0.01 1	1 2	μA
Output overvoltage protection		-	21	23.5	25	V
Over temperature protection			135	150	165	°C
Over temperature hysteresis				10		°C

**Note 1:** Internal supply voltage at VDD pin is only for internal circuit use. It should not be loaded externally. **Note 2:** Digital inputs have active pull-down by 410k $\Omega$  which is disabled to save current when inputs are pulled high. TBD = To Be Defined

# THERMAL INFORMATION

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Junction to ambient thermal resistance	Reja	Note 1		26.6		°C/W

Note 1: Simulated value for Jedec JESD51-5 4-layer 1.6mm thick board with 4x4=16 pcs thermal vias under QFN package thermal pad.



#### **APPLICATION DIAGRAMS**



Figure 1. MAS6230 application diagram with fixed 5V output

The MAS6230 can be kept enabled always by connecting the EN pin to the VDD pin since the internal supply voltage is available from the VDD pin also in shut down mode.







# **APPLICATION DIAGRAMS (continued)**



Figure 3. MAS6230 application diagram with regulated MCU selecting between four output voltage options



#### **DETAILED DESCRIPTION**

MAS6230 is a high efficiency synchronous buckboost DC/DC converter with both wide input and output voltage range. Four integrated low resistance N-channel power transistors minimize application solution size and maximize efficiency.

The device is controlled on and off using single enable pin (EN). In shut-down (EN=LOW) the DC/DC converter is turned off and only a small quiescent current is drawn from the input ( $V_{IN}$ ).

When enabled (EN=HIGH) the DC/DC converter is on and selection between buck ( $V_{OUT} < V_{IN}$ ) and boost ( $V_{OUT} > V_{IN}$ ) operating modes is done automatically. There is hysteresis between buck and boost operating modes which makes the DC/DC converter stable with wide range of external components. The undervoltage lockout (UVLO) circuit prevents the DC/DC converter to operate at too low input voltage conditions. The device enters shut-down if input voltage drops below typ 6.3V independent of enable pin state. When enabled the device is on if input voltage has risen above typ 6.6V.

MAS6230 regulates output voltage (V<sub>OUT</sub>) to level which is adjusted by feedback voltage (VFB) and two feedback resistor values (RFB1, RFB2). See application figures 1-3. The feedback voltage can be selected from four options using FBSEL0-1 pins. This makes it possible to output four different output voltages which are selectable just by FBSEL0-1 pin control as shown in the figures 2-3. See also Output voltage adjustment in the Application information section.

#### **APPLICATION INFORMATION**

#### Inductor selection

Inductor needs to be rated typically to 6A or higher current. The dc-resistance of inductor has contribution to power losses and to achieve high efficiency and output power the dc-resistance should be minimized. Table 1 shows two examples of inductors. The larger toroid coil with low dc-resistance is recommended for highest efficiency and highest output power applications. The smaller SMD inductor can be used in more space-constrained applications with lower output power requirement.

 Table 1. Inductor examples

Part number	Manuf.	Package	L [µH]	$R_{DC}$ [m $\Omega$ ]	IRATED [A]
DTMSS-12.5	Feryster	Toroid core Ø 12.5mm	6.8	7.22	8
XAL5050-682ME	Coilfraft	SMD 5x5x5.1mm	6.8	26.75	6

#### **Capacitor selection**

All capacitors must be low loss (low ESR) ceramic capacitors. Recommended capacitors values are shown in the table 2. The CFWD=47nF capacitor value is suited to be used with feedback resistor values RFB1=300 $\Omega$  and RFB2=4.7k $\Omega$ .

Capacitor	Nominal value	Voltage rating	Туре
CIN	22µF	50V	Ceramic MLCC capacitor
COUT1	47μF	25V	Ceramic MLCC capacitor
COUT2	100μF	25V	Organic polymer capacitor
CVDD1	22μF	10V	Ceramic MLCC capacitor
CVDD2	4.7μF	10V	Ceramic MLCC capacitor
CFLY1	100nF	10V	Ceramic MLCC capacitor
CFLY2	100nF	10V	Ceramic MLCC capacitor
CFWD	47nF	25V	Ceramic MLCC capacitor

#### Table 2. Recommended capacitor values



### **APPLICATION INFORMATION (continued)**

#### Output voltage adjustment

MAS6230 produces regulated output voltage (VOUT) which depends on selected feedback voltage option (VFB) and feedback resistor values (RFB1, RFB2) as described in the equation 1. See also figures 1 and 2 for the feedback resistor configuration.

$$VOUT = VFB \cdot \left(1 + \frac{RFB2}{RFB1}\right)$$
 Equation 1.

There are four different feedback voltage (VFB) options selectable by the FBSEL0-1 pins. See table 3.

FBSEL1	FBSEL0	VFB [V]
GND	GND	1.2
GND	VDD	0.9
VDD	GND	0.54
VDD	VDD	0.3

 Table 3. Feedback voltage selection options

For example by choosing RFB2=4.7k $\Omega$  and RFB1=300 $\Omega$  the four standard output voltages 20V, 15V, 9V and 5V are selectable using the FBSEL0-1 pins.

 $\label{eq:FBSEL1/0=GND/GND: V_{OUT}=1.2*(1+4.7k\Omega / 300\Omega) = 20V \\ FBSEL1/0=GND/VDD: V_{OUT}=0.9*(1+4.7k\Omega / 300\Omega) = 15V \\ FBSEL1/0=VDD/GND: V_{OUT}=0.54*(1+4.7k\Omega / 300\Omega) = 9V \\ FBSEL1/0=VDD/VDD: V_{OUT}=0.3*(1+4.7k\Omega / 300\Omega) = 5V \\ \end{array}$ 

#### Maximum output current (typical)

Maximum output current is limited mainly by heating effect. The power losses should be kept below max 2.5W to keep operating junction temperature below max +125°C when ambient temperature is at max +60C°. Second output current limitation comes from coil current limiting of MAS6230. This output current limitation can take place in boost operating mode ( $V_{IN} < V_{OUT}$ ) at which input current is increased. Figure 4 illustrates maximum output currents (IOUT[A] max) in typical case at different input voltages (VIN[V]) and at four output voltage options 5V, 9V, 15V and 20V.



Figure 4. Typical MAS6230 maximum output currents (I<sub>OUT</sub> max) at different input (V<sub>IN</sub>) and output (V<sub>OUT</sub>) voltages



#### THERMAL DESIGN CONSIDERATIONS

MAS6230 is a high power device which can output up to 60W power to load. It has high efficiency but at maximum output loading the power losses can be over two watts. This puts special attention to printed circuit board (PCB) thermal design to avoid device overheating.

Heat transfer capability from silicon die to ambient is described by thermal resistance ( $R_{\theta JA}$ ) which is ratio of temperature difference ( $\Delta T_{JA}$ ) between silicon junction and ambient and power loss ( $P_{\text{loss}}$ ). See equation 2.

$$R_{\theta JA[^{\circ}C/W]} = \frac{\Delta T}{P_{loss}} = \frac{T_J - T_A}{P_{loss}}$$

#### Equation 2.

Thermal resistance must be low enough to keep the silicon junction temperature below maximum specified operating junction temperature at maximum ambient temperature and power loss conditions. Using equation 2 we can calculate maximum thermal resistance for conditions  $T_J(max)=+125^{\circ}C$ ,  $T_A(max)=+60^{\circ}C$  and  $P_{loss}(max)=2.5W$ .

$$R_{\theta JAmax[^{\circ}C/W]} = \frac{125^{\circ}C - 60^{\circ}C}{2.5W} = 26^{\circ}C/W$$

Thermal resistance is characteristic of package and thermal design of the PCB. The QFN package contains exposed thermal pad which offers an effective heat transfer path from the silicon die to the PCB. On PCB the thermal heat transfer can be considered taking place only via metal layers since copper has 400-1400 times larger thermal conductivity than the insulating FR-4 substrate. To achieve sufficiently low thermal resistance it is necessary to use thermal vias under the exposed thermal pad and to connect the vias to inner and other available layer continuous metal planes on a multi-layer board. In thermal design the key point is to transfer the heat from component to continuous metal planes which spread the heat to wider PCB area wherefrom it can further transfer to ambient.



Figure 5. QFN 7x7 package thermal via design illustration on a 4-layer board

Figure 5 illustrates QFN 7x7 package thermal via design on a multi-layer board. Board and via parameters are listed in table 3.

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Parameter	Unit	Value	Note
Finished PCB thickness	mm	1.6	
Layers	pcs	4	
Copper thicknesses			As many as possible continuous GND planes
Top/Bottom	oz (μm)	2 (70)	connected to QFN package exposed thermal
Inner layers		1 (35)	pad using thermal vias
PCB thermal landing	mm	5.3x5.3	Same as QFN package exposed pad size
Thermal via			Solid thermal vias
Diameter	mm	0.3	
Pitch		1	
Thermal via configuration	pcs	4x4=16 or 5x5=25	



# **TYPICAL ELECTRICAL AND THERMAL CHARACTERISTICS**

# Output voltage stability (typical)



Figure 6. Output voltage VOUT[V] vs. output current IOUT[A] at VOUT=5V selection and at different VIN[V]



Figure 7. Output voltage VOUT[V] vs. output current IOUT[A] at VOUT=9V selection and at different VIN[V]





Figure 8. Output voltage VOUT[V] vs. output current IOUT[A] at VOUT=15V selection and at different VIN[V]



Figure 9. Output voltage VOUT[V] vs. output current IOUT[A] at VOUT=20V selection and at different VIN[V]



#### Efficiency (typical)

The efficiency figures 10-13 include also two calculated efficiency curves (dashed lines) corresponding to 2W and 2.5W power losses. Required minimum efficiency ( $\eta$  min) for maximum power loss (Ploss max) can be calculated from equation 3.

 $\eta_{min}[\%] = 100\% \cdot \frac{P_{out}}{P_{loss\,max} + P_{out}} = \frac{100\%}{\frac{P_{loss\,max}}{P_{out}} + 1} = \frac{100\%}{\frac{P_{loss\,max}}{V_{out} \cdot I_{out}}}$ Equation 3.

For example if maximum allowed power loss is Ploss max =2.5W the minimum required efficiency at VOUT=5V, IOUT=3A (POUT=15W) would be  $\eta$  min =100%/[2.5W/(5V\*3A)+1]=85.7%.



Figure 10. Efficiency n[%] vs. output current IOUT[A] at VOUT=5V selection and at different VIN[V]



Figure 11. Efficiency  $\eta$ [%] vs. output current IOUT[A] at VOUT=9V selection and at different VIN[V]





Figure 12. Efficiency n[%] vs. output current IOUT[A] at VOUT=15V selection and at different VIN[V]



Figure 13. Efficiency n[%] vs. output current IOUT[A] at VOUT=20V selection and at different VIN[V]



#### Power loss (heating power) (typical)

Figures 14-17 present power losses at different input voltage VIN[V] and output current IOUT[V] conditions when output voltage VOUT[V] voltage is 5V, 9V, 15V and 20V. The dark green contour map represents power losses up to 2W and orange countour map corresponds to losses up to 2.5W.



Figure 14. Power loss Ploss[W] vs. output current IOUT[A] at VOUT=5V selection and at different VIN[V]



Figure 15. Power loss Ploss[W] vs. output current IOUT[A] at VOUT=9V selection and at different VIN[V]





Figure 16. Power loss Ploss[W] vs. output current IOUT[A] at VOUT=15V selection and at different VIN[V]



Figure 17. Power loss Ploss[W] vs. output current IOUT[A] at VOUT=20V selection and at different VIN[V]



#### **DEVICE OUTLINE CONFIGURATION**





# QFN-48 7x7x0.75 PIN DESCRIPTION

Parameter	Unit	Value	Note	
Pin Name	Pin	Туре	Function	Note
-	1	-	Not connected	1
GND	2-3	G	Ground	
FBSEL1	4	DI	Output voltage selection bit 1	
FBSEL0	5	DI	Output voltage selection bit 0	
EN	6	DI	DC/DC enable	
GND	7	G	Ground	
FLY1A	8	AO	Flying capacitor 1 positive terminal	
FLY1B	9	AO	Flying capacitor 1 negative terminal	
GND	10-11	G	Ground	
-	12-13	-	Not connected	1
GND	14	G	Ground	
VIN	15-17	Р	DC/DC Input voltage	
VDD	18	AO	Internal power supply	
-	19	-	Not connected	1
VOUT	20-21	AO	DC/DC output voltage	
GND	22-23	G	Ground	
FB	24	AI	Feedback	
-	25	-	Not connected	1
GND	26-27	G	Ground	
FLY2B	28	AO	Flying capacitor 2 negative terminal	
FLY2A	29	AO	Flying capacitor 2 positive terminal	
GND	30	G	Ground	
TM	31	DIO	Test mode pin	2
FORCEBUCK	32	DI	Force Buck selection	3
			Buck operation: FORCEBUCK=VDD	
			Buck/boost operation: FORCEBUCK=GND	
GND	33-35	G	Ground	
-	36-37	-	Not connected	1
GND	38	G	Ground	
SW2	39-41	AO	Inductor switch terminal 2	
-	42	-	Not connected	1
SW1	43-45	AO	Inductor switch terminal 1	
MEMO	46	AI	Memory test input	2
GND	47	G	Ground	
-	48	-	Not connected	1

G = Ground, P = Power, D = Digital, A = Analog, I = Input, O = Output

Note 1: Not connected pins should be connected to ground

Note 2: The TM and MEMO pins are only for internal testing purpose and must be connected to GND in the application

**Note 3:** In applications which require only but normal team ( $V_{IN} > V_{OUT}$ ) the FORCEBUCK is connected to VDD pin. This forces device into buck operation by disabling boost operation ( $V_{IN} < V_{OUT}$ ). In other applications the FORCEBUCK is connected to GND. This enables both buck and boost operations.

Note: Thermal pad (exposed pad) must be connected to continuous ground plane(s) using thermal vias to guarantee maximal heat transfer



# PACKAGE (QFN-48 7x7x0.75) OUTLINE



BOTTOM VIEW



# PACKAGE (QFN-48 7x7x0.75) OUTLINE (continued)

Symbol	Min	Nom	Max	Unit
		PACKAGE DIMENSIC	NS	
A	0.7	0.75	0.8	mm
A1	0	0.035	0.05	mm
A2		0.55	0.57	mm
A3	0.203 REF			mm
b	0.2	0.25	0.3	mm
D		7 BSC		
E		7 BSC		
е		0.5 BSC		
J (Exposed.pad)	5.2	5.3	5.4	mm
K (Exposed.pad)	5.2	5.3	5.4	mm
Ĺ	0.35	0.4	0.45	mm



#### ORDERING INFORMATION

Product Code	Product	Package	Comments
MAS6230BA1Q2406	High Efficiency 60W Buck-Boost	QFN-48 7x7x0.75, Pb-	Tape and Reel
	DC/DC Converter	free, RoHS compliant	3000 pcs / r

#### LOCAL DISTRIBUTOR

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